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Closely-spaced double-row microstrip RF arrays for parallel MR imaging at ultrahigh fields

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Abstract

Radiofrequency (RF) coil arrays with high count of elements, e.g., closely-spaced multi-row arrays, exhibit superior parallel imaging performance in MRI. However, it is technically challenging and time-consuming to build multi-row arrays due to complex coupling issues. This paper presents a novel and simple method for closely-spaced multi-row RF array designs. Induced current elimination (ICE) decoupling method has shown the capability of reducing coupling between microstrip elements from different rows. In this study, its capability for decoupling array elements from the same row was investigated and validated by bench tests, with an isolation improvement from -8.9 dB to -20.7 dB. Based on this feature, a closely-spaced double-row microstrip array with 16 elements was built at 7T. S_{21} between any two elements of the 16-channel closely-spaced was better than -14 dB. In addition, its feasibility and performance was validated by MRI experiments. No significant image reconstruction-related noise amplifications were observed for parallel imaging even when reduced factor (R) achieves 4. The experimental results demonstrated that the proposed design might be a simple and efficient approach in fabricating closely-spaced multi-row RF arrays.

Keywords

Microstrip; RF array; induced current elimination (ICE); decouple; parallel imaging; magnetic resonance image (MRI)

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Introduction

Since its introduction, ultrahigh field (UHF) MRI, i.e., 7T and beyond, has received a great deal of attention in in vivo MR imaging studies due to its increased signal-to-noise ratio (SNR) and imaging contrast [1–3]. However, at the frequency of 300 MHz (Larmor frequency of proton at 7T) or higher, the wavelength in biological tissue is significant shortened [4–6]. Therefore, it is technically challenging to design large-sized radiofrequency (RF) coil. A variety of RF array coils has been developed in UHF MRI, including traditional L/C loop coil [7–10], microstrip coil [11–18] and radiative coil [19–21]. Due to its simple structure and low radiation loss, microstrip technique has been a popular choice in designing RF array in UHF MRI.

Recently, a multi-row microstrip array using induced current elimination (ICE)/magnetic wall decoupling was proposed and realized at 7T [22]. In the ICE-decoupled multi-row array, a “figure of 8”-shaped resonator (referred to as the decoupling element) was used to suppress the EM coupling from adjacent elements of different rows. The self-impedance of the decoupling element (X_{dd}) was adjusted to meet the decoupling condition: $X_{dd}=X_{cd}^2/X_{cc}$, where X_{cd} is mutual impedance of coil element and decoupling element and X_{cc} is the self-impedance of adjacent elements from different rows. It has been demonstrated that the decoupling element acts like a “magnetic wall” and exhibits a “shielding effect” [23, 24]. This shielding effect might also be beneficial to reduce the coupling between adjacent elements from the same row. In other words, the self-impedance of the decoupling element (X_{dd}), which is intended to meet the decoupling condition of coil elements from adjacent rows, might also meet the decoupling condition of adjacent elements from the same row to a certain degree.

To validate this assumption, a double-row microstrip array with two blocks was built and comparatively investigated. Each block consisted of two microstrip elements from different rows and a decoupling element. Both the experimental and simulation results showed that the decoupling element, which was used for reducing the coupling of elements from different rows, can also reduce the coupling of adjacent elements from the same row. Therefore, the ICE decoupling method provides a simple and efficient approach to building closely-spaced double-row microstrip arrays. In this study, a 16-channel double-row microstrip array with coil elements closely placed was built at 7T. Bench test and MR imaging results were obtained using this array to demonstrate its feasibility and parallel imaging performance [25–28].

Materials and Methods

A) Comparative analysis of ICE decoupling for elements from the same row

Fig. 1A shows the photograph of an ICE-decoupled double-row microstrip array with two blocks. Each block consists of two microstrip elements (9 cm × 4 cm × 1.5 cm in dimension) and a decoupling element (21 cm × 1 cm × 1.5 cm in dimension). The distance between two blocks is 2 cm.

The microstrip element is a typical straight microstrip coil made of a Teflon bar and copper sheets [11]. Two capacitors were terminated at both ends of the strip conductor. Width of strip conductors and ground plates were 1 cm and 4 cm, respectively. The decoupling element is a resonator with two capacitors terminated at both ends and one variable capacitor terminated at the center (Fig. 1C). In each block, the decoupling element was closely placed to the coil elements, with a gap of about 2 mm. For comparison, a two-channel microstrip coil array with the same parameters and dimensions was constructed, as shown in Figs. 1B and 1D.

The decoupling performance among any two elements was evaluated by transmission coefficient (S_{21}). All coil elements were matched to 50 ohm and tuned to 297.2 MHz, which is the Larmor frequency of the utilized 7T MRI system. Before two blocks of the double-row array were placed together, pairs of each block, i.e., element 1-element 3 and element 2-element 4, were tuned, matched and decoupled separately. RF cable traps were placed between cables and coil elements to avoid the “cable resonance”.

Fig. 2A shows S_{11} and S_{21} plots of two elements of the single-row microstrip array (Fig. 1B). The coupling between two elements is -8.9 dB. This is reasonable given that the thickness of coil elements achieves 1.5 cm while the space between them is only 2 cm. Fig. 2C shows the S_{21} plot of elements from different rows, e.g., elements 1 and 3 in Fig. 1A. As described above, the decoupling elements were tuned to obtain best isolation between elements from different rows. Therefore, their coupling was sufficiently small (-26 dB) and the dip of S_{21} plot was exactly at 297.2 MHz.

Fig. 2B shows the S_{11} and S_{21} plots of two elements of the same row, e.g., elements 1 and 2 in Fig. 1A. It was found that the isolation could still achieve -20.7 dB even though the dip of the S_{21} plot deviates from 297.2 MHz. These results indicate that the decoupling element can reduce the coupling of adjacent elements from the same row as well as those from different rows.

To further demonstrate the shielding effect of the decoupling element, magnetic field distributions of a microstrip element with and without the decoupling element were evaluated by a full-wave EM simulation tool (HFSS, ANSYS, Canonsburg, PA, US). Dimensions of the decoupling element and coil elements were exactly the same as those in practice. Manual mesh definition was used in a 3-D domain to accelerate simulation convergence and the convergence condition was set as: $s < 0.005$ to ensure reliability. Values of all capacitors were obtained by a circuit co-simulation approach [29, 30]. In the EM simulation, each element was matched to 50 ohm ($S_{11} < -30$ dB) and the two elements from the ICE-decoupled double-row block were decoupled well ($S_{21} < -25$ dB).

Fig. 3A shows the magnetic field distribution of a single microstrip element without the decoupling element. Fig. 3B shows the magnetic field distribution of a microstrip element from the ICE-decoupled double-row block. The magnetic field was normalized to 1 W input power. Due to the symmetrical structure, the single microstrip element has a symmetrical magnetic field distribution, as shown in Fig. 3A. For the microstrip element from the ICE-decoupled double-row block, however, its magnetic field near the decoupling element was

obviously diminished, which might be attributed to the induced current of the decoupling element. This diminution makes the magnetic field more confined and thus could improve the isolation between adjacent elements in the same row, which has been validated by S-parameter results described above.

B) Volume-typed double-row microstrip array with 16 elements

To test the feasibility of closely-spaced double-row design for volume-typed microstrip array, up to 16 channels, i.e., 8 blocks, were built in practice. These blocks were attached onto a soft canvas and mounted on a cylindrical acrylic former (16 cm in diameter), as shown in Figs. 4A and 4B. It is worth noting that the ICE-decoupled double-row microstrip array does not need physical connections between adjacent blocks, which make it possible and practical for flexible array designs. The space of two adjacent block is only about 1.3 cm. Scattering (S-) parameter matrix of the 16-channel array was measured with an Agilent 5071C network analyzer. All other coil elements were terminated with 50- Ω loads when any two elements were measured. The S_{II} plots were also used to calculate Q values.

A homogenous bottle-shaped phantom (12 cm in diameter and 22 cm in height) filled with 2.5 g NiSO_4 and 15 g NaCl was used to test the performance of the closely spaced double-row array. The EM parameters of the phantom were measured by a dielectric probe (DAK-12, SPEAG, Switzerland): conductivity $\sigma=0.59$ S/m; relative permittivity $\epsilon_r=78$. MR imaging of the phantom was performed on a whole-body MRI scanner (7T MAGNETOM, Siemens Healthcare, Erlangen, Germany). GRE sequences were used and the scan parameters were as follows: flip angle (FA)= 25 degree, TR=100 ms, TE=10 ms, field of view (FOV)= 210 $\times$ 210 mm², matrix= 256 $\times$ 256, slice thickness=5 mm, number of excitation (NEX)=1, bandwidth= 320 Hz/pixel. Accelerated and residual GRE images with reduced factor (R) of 2, 3 and 4 were obtained to show the parallel imaging capability of the 16-channel array. The accelerated images were reconstructed with generalized auto calibrating partially parallel acquisitions (GRAPPA) method. In the reconstruction, 24 auto-calibration signal (ACS) lines were used for each reduced rate. MRI scans were conducted two times given that our RF interface can only operate 8 channels at one time. For each scan, 8 channels from one row were connected to the RF interface with a phase increment of 45 degree between adjacent channels. Meanwhile, the other channels were terminated with 50- Ω loads. During the two scans, the positioning of a subject was carefully kept unchanged via the line markers made on the patient table.

Results

A) Bench test results of the double-row microstrip array with 16 elements

Figs. 5A and 5B show the S-parameter matrix and noise covariance matrix of the 16-channel array, respectively. The coil numbering in Fig. 5 corresponded to that in Fig. 4A. The mean S_{21} was -19.2 dB between the adjacent elements from different rows, e.g., elements 1 and 8. The mean S_{21} was -16.1 dB between adjacent element from the same row, e.g., elements 1 and 2. From our practical experience, the decouple element make the cross talk between diagonal elements (e.g., elements 1 and 10) even a little stronger. However, it still achieves better than -14 dB in the loaded case, as shown in Fig. 5A. The isolation could be further

improved by optimizing the thickness of the Teflon substrate. The noise covariance results (Fig. 5B) were in agreement with the S-parameter results. Given that MR experiments were conducted using the two rows separately (each row at one time), the noise covariance between elements from different rows was almost zero. The unloaded Q values were approximately 205 and the Q values loaded with the phantom were in the range of 102~114.

B) MR imaging results of the closely-spaced double-row microstrip array with 16 elements

Fig. 6 shows the individual images in the transverse and sagittal planes from each coil element and their combination. These images were reconstructed from raw data and combined together with root-sum-of-squares (RSS) method. These images indicated that the ICE-decoupled double-row microstrip array design is feasible in designing close-spaced volume-typed arrays for MR imaging at the ultrahigh field of 7T. To evaluate the parallel imaging performance, sagittal GRE images acceleration of 2 $\times$, 3 $\times$ and 4 $\times$ were shown in Fig. 7. No significant image reconstruction-related noise amplifications were observed for parallel imaging even when reduced factor (R) achieves 4. The parallel imaging results were also consistent with the S-parameter and noise covariance results described above.

Discussions and Conclusion

Most of the multi-row transceiver arrays are based on traditional L/C loop resonators. For example, Avdievich [9] and Shajan et al. [31] developed inductively decoupled double-row loop arrays for human brain imaging at 7T and 9.4T, respectively; Thalhammer et al. [32] used capacitively decoupled double-row loop arrays for cardiac imaging at 7T. Compared with single-row RF arrays, these double-row arrays improve longitudinal imaging coverage, homogeneity and SNR [9]. For both inductively decoupled and capacitively decoupled multi-row loop arrays, however, it is technically challenging and time-consuming to fabricate since both the coupling of elements from different rows and from the same row have to be reduced.

In this study, we aim to find a simple and efficient way of building closely-spaced multi-row arrays for parallel MR imaging. Microstrip line resonators were chosen as the basic coil elements and ICE/magnetic wall method was used for decoupling. ICE/magnetic wall decoupling has proven to be a general decoupling method, and has been successfully applied for almost all kinds of RF coils in MRI, including L/C loop, microstrip line, monopole and dipole coils [23, 24, 33, 34]. In previous ICE-decoupled RF array designs, decoupling elements were placed between two coupled coil elements. To reduce both the coupling from the same row and that from different rows, in this study, the decoupling element was placed next to the coil elements rather than inserted between them. In this case, the self-impedance of the decoupling element (X_{dd}), which is intended to meet the decoupling condition of coil elements from adjacent rows, might also meet the decoupling condition of adjacent elements from the same row to a certain degree. Experimental results demonstrated that the ICE decoupling method could also reduce the coupling of adjacent elements in the same row, with an isolation improvement from -8.9 dB to -20.7 dB (Fig. 2).

Given that the ICE decoupling method could reduce both the coupling from different rows and from the same row for microstrip arrays, it is much less laborious to fabricate closely-

spaced multi-row microstrip arrays. We constructed and investigated a volume-typed microstrip array with 2×8 elements. Decoupling elements were tuned to obtain best isolation between adjacent elements of each block (i.e., adjacent elements of different rows). When 8 blocks were placed together, the isolation of adjacent elements from the same row could still be better than -15 dB without retuning decoupling elements. The excellent decoupling performance of the 16-ch array has also been validated by noise covariance matrix and parallel imaging results. It is worth noting that the “figure of 8”-shaped decoupling element makes the cross talk between diagonal elements a little stronger. Thus the diagonal elements show the strongest coupling, as shown in Fig 3. Further studies are needed to reduce the coupling between diagonal elements.

It has previously been shown that multi-row transceiver arrays exhibit the capability of RF shimming and pTx along z-direction [35, 36], which is beneficial in providing more uniform transmit field. Due to the limitation of our RF interface, two rows of the 16-channel array cannot be excited simultaneously. Therefore, 3D RF shimming capability of the 16-channel array was not shown in this work. However, based on the S-parameter and MR imaging results, we believe this double-row closely-spaced transceiver array could be an efficient tool to perform 3D RF shimming or pTx.

In summary, we proposed and developed a closely-spaced double-row microstrip transceiver array for parallel MR imaging at the ultrahigh field of 7T. Its feasibility and performance was validated by bench tests and MRI experiments. This work has also paved the way for building multi-row microstrip transceiver arrays for human head or body imaging with higher channel count, e.g. 32-channel.

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Fig. 1. Photographs and schematic diagram of a double-row microstrip array with two blocks (**A** and **C**) and a single-row microstrip array with two elements (**B** and **D**).

Fig. 2.

A: S_{11} and S_{21} plots of two elements of the conventional single-row microstrip array, i.e., elements 1 and 2 (Fig. 1B). **B:** S_{11} and S_{21} plots of two elements of the same row, e.g., elements 1 and 2 (Fig. 1A). **C:** S_{21} plot of elements from different rows, e.g., elements 1 and 3 (Fig. 1A).

Fig. 3. Magnetic field distributions of a single microstrip element (**A**) and a coil element from the ICE-decoupled double-row block (**B**) in the transverse plane. Dimensions of the decoupling element and coil elements were exactly the same with that in practice.

Fig. 4. Coil element arrangement (**A**) and photograph (**B**) of the 16-channel double row microstrip array. Coil elements were closely placed along the surface of a cylindrical former (16 cm in diameter).

Fig. 5. S-parameter matrix (**A**) and noise covariance matrix (**B**) of the 16-channel double-row microstrip array. The isolation between any two coil elements is better than -14 dB.

Fig. 6.

(**A**) Combined GRE image and sub-images from each coil element in the transverse plane with parameters: FA=25 degree, TR=100 ms, TE=10 ms, FOV= 210×210 mm², matrix= 256×256, slice thickness=5 mm, NEX=1, bandwidth= 320 Hz/pixel. (**B**) Combined GRE image and sub-images from each coil element in the sagittal plane using the same sequence.

Fig. 7. Accelerated GRE images (top row) and residual images (bottom row) with reduced factor (R) of 1 (no acceleration), 2, 3 and 4 in the sagittal plane.